

# Lecture 16

## PID controller

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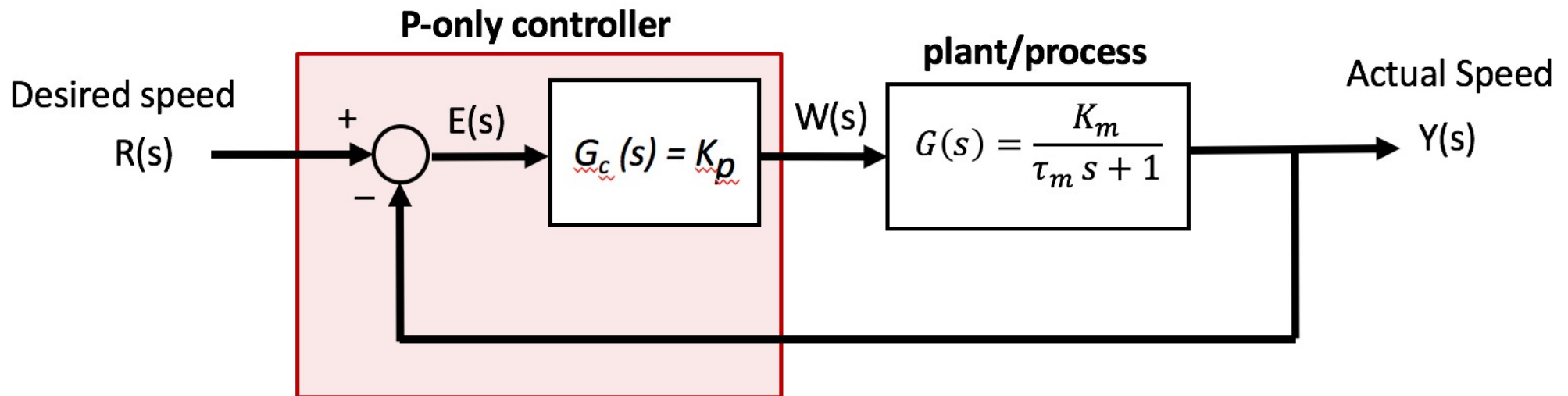
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# Limitations of Proportional-only (P) control

- ◆ In **proportional-only** control, the controller output is given by:  $w(t) = K_p e(t)$
- ◆ Using P-only control is simple, but often insufficient because:
  1. If  $K_p$  is small, **error**  $e(t)$  can be large (i.e. there is an offset error between set-point and controlled output variable, in this case, speed of motor).
  2. If  $K_p$  is large, the system may oscillate (i.e. become unstable).
  3. Even if the system is stable, it may take a long time to settle to its final output value or exhibition large overshoots.
  4. It may not have sufficient tolerance to perturbations or disturbances.

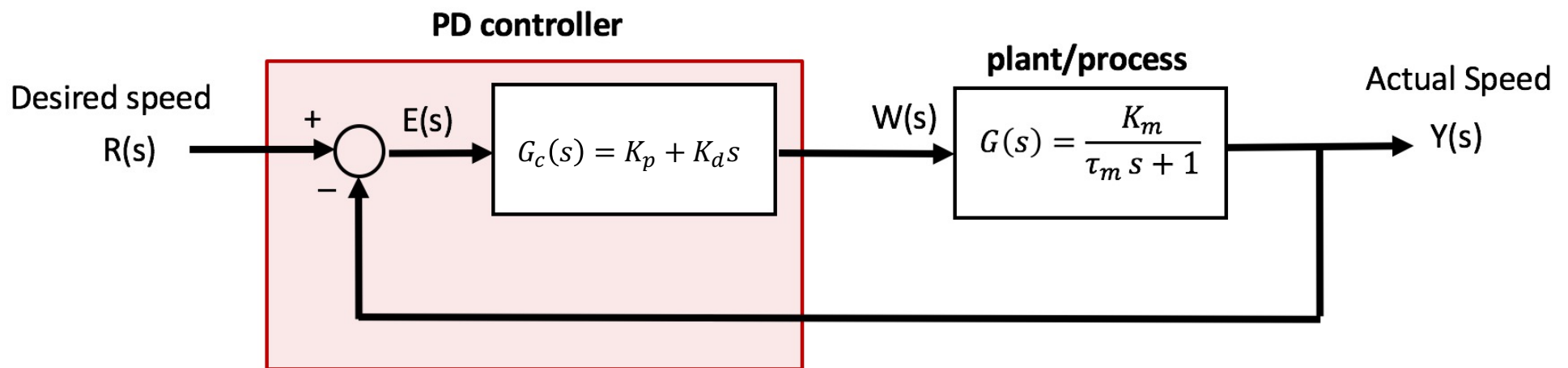


# Proportional - Derivative (PD) Control

- ◆ We can add another term to include the rate of change of the error  $\dot{e}(t)$ . This is known as a **proportional-derivative** (PD) controller:  $w(t) = K_p e(t) + K_d \dot{e}(t)$
- ◆ In computers, the **derivative term**  $\dot{e}(t)$  is usually calculated by taking the **difference** between current error value  $e[n]$  and the previous error value  $e[n-1]$ :

$$\text{differential term at time } n = K_d (e[n] - e[n - 1]) / \Delta t$$

- ◆ The main advantages of the PD controllers are:
  1. It can reduce the overshoot of a proportional-only controller response because PD controller takes into account the rate of change in error.
  2. It can also improve the system's tolerance to external disturbances.



# Proportional - Integral (PI) Control

- ◆ Alternatively, we can add an **integral term** to the controller. This is known as a PI controller:

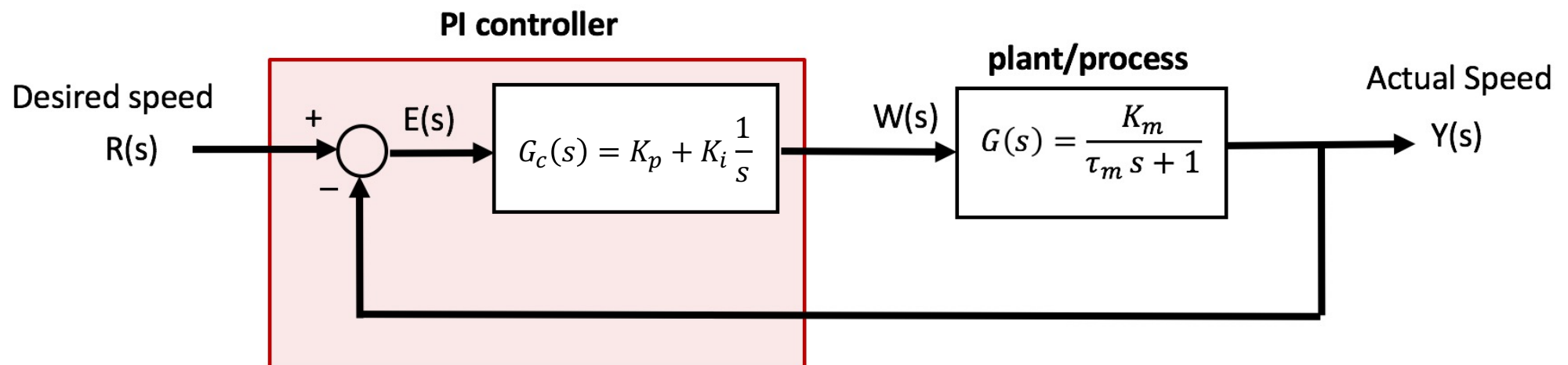
$$u(t) = K_p e(t) + K_i \int e(\tau) d\tau$$

- ◆ The integral term is implemented on a digital microprocessor as **summation** ( $\Delta t$  is the sampling interval):

$$\text{integral term at time } n = K_i \Delta t \sum_{k=0}^n e[n - k]$$

- ◆ The main advantages of the PI controller are:

1. It eliminates **steady-state error**.
2. It can help with **stability of the system**, especially if  $K_p$  is large.



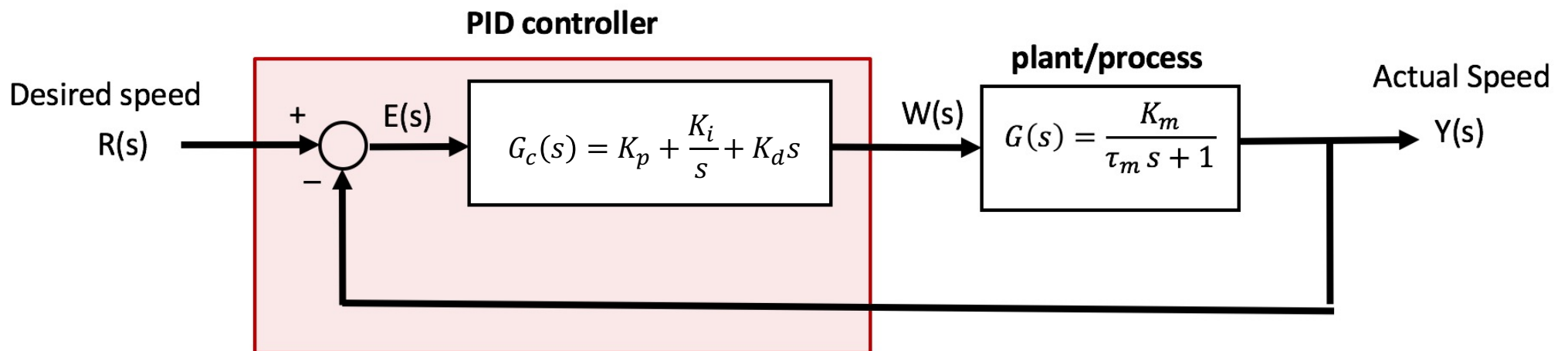
# PID Control

- ◆ Finally, we can combine all three terms to form a PID controller:

$$u(t) = K_p e(t) + K_i \int e(\tau) d\tau + K_d \dot{e}(t)$$

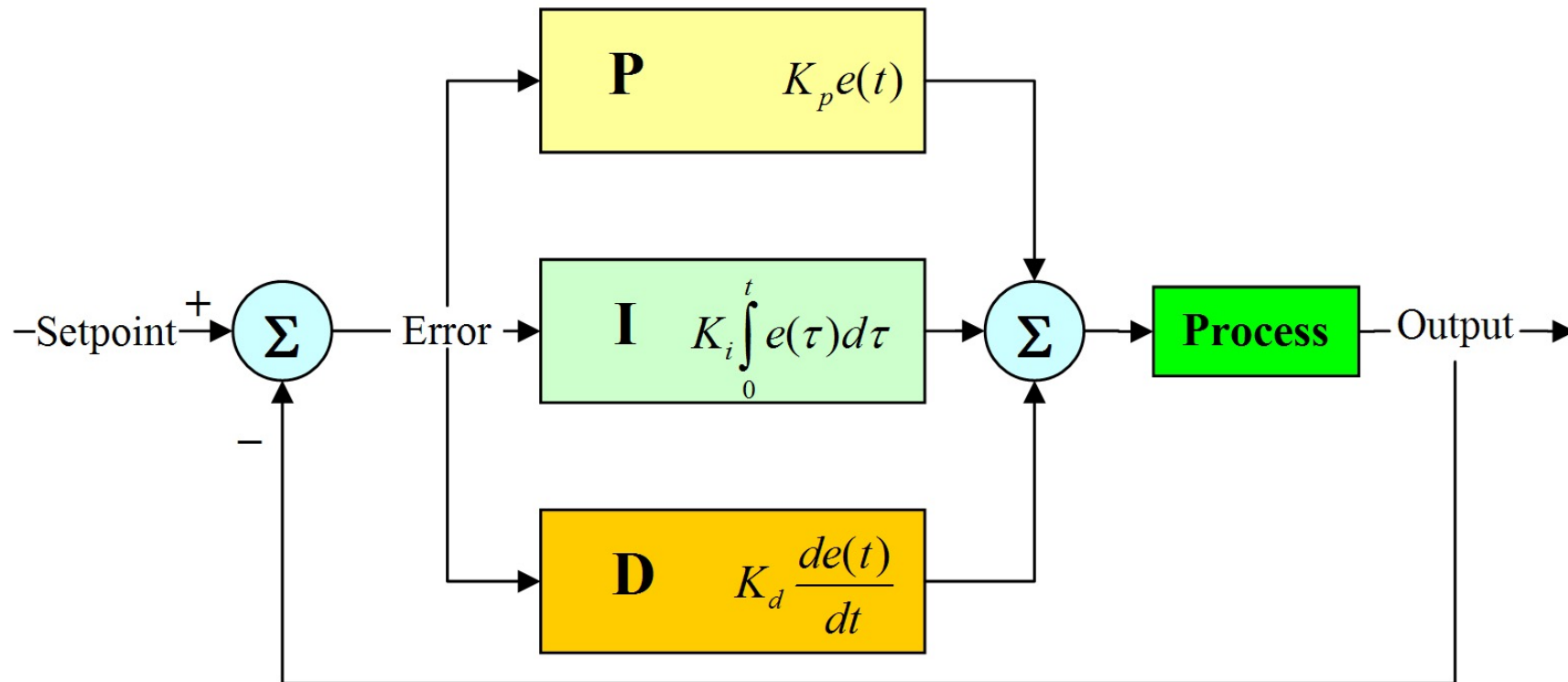
- ◆ This has the advantages of **ALL** three types of feedback control (P, I and D):

1. **Removal** of steady-state **error** due to I.
2. **Reduce** the amount of **overshoots** (due to be I and D).
3. **Improve** the **transient** response to make it faster (due to both I and D).
4. **Improve stability** of the system.
5. Better perturbation tolerance.

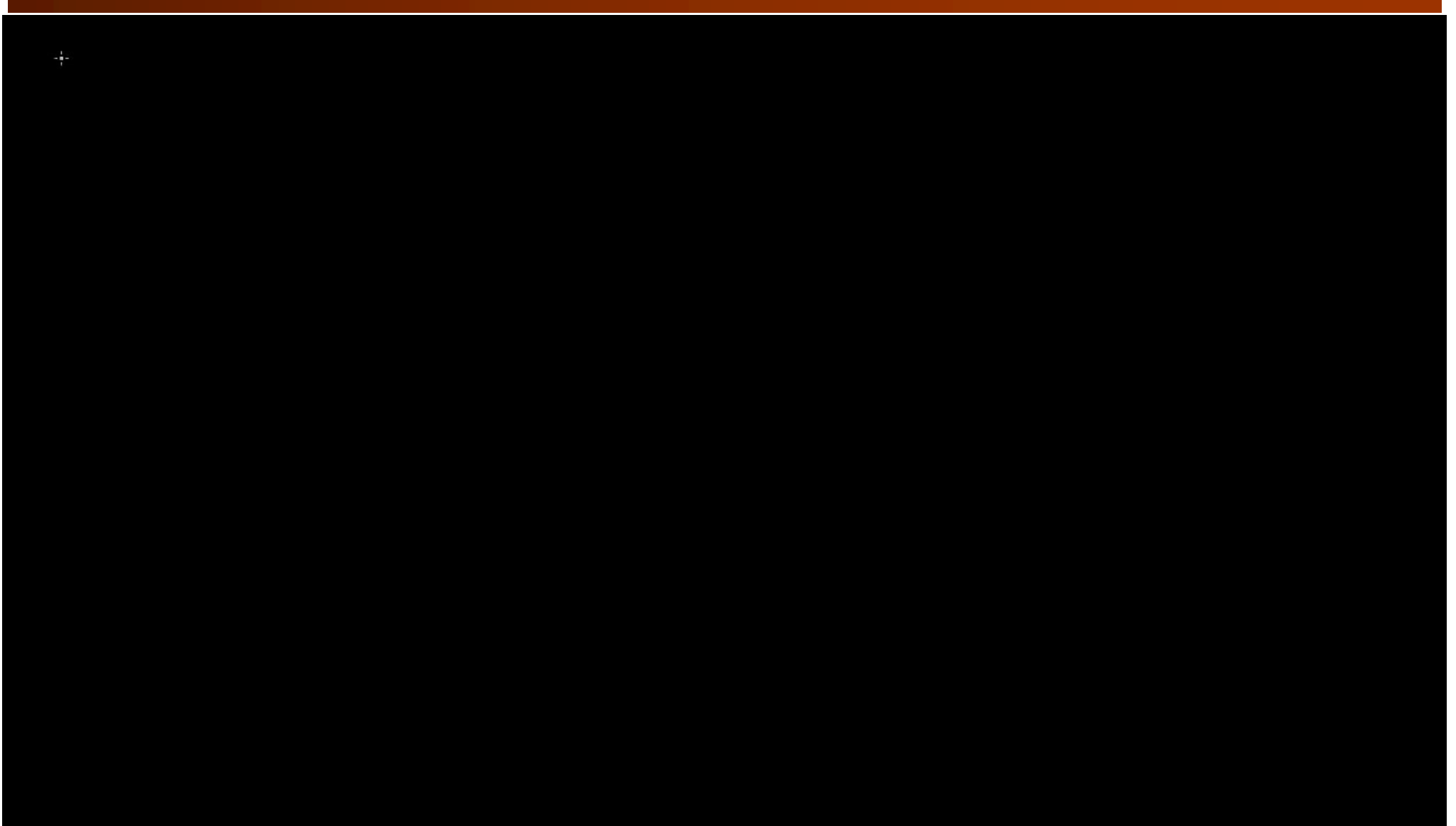


# PID Controller block diagram

- ◆ Here is a schematic of an implementation of a typical PID controller.
- ◆ In practice, we often use PI (e.g. driving a motor), or PD (e.g. balancing two-wheel vehicle), and NOT all three terms.



# Use PID controller to drive a car



# Tuning of a PID Controller

- ◆ Choosing the correct values for  $K_p$ ,  $K_d$  and  $K_i$  is known as **tuning** the controller.
- ◆ **Impact of various gains** on step response of a system:

PID Gain	Percent Overshoot	Settling Time	Steady-State Error
Increasing $K_p$	Increases	Minimal impact	Decreases
Increasing $K_i$	Increases	Increases	Zero steady-state error
Increasing $K_D$	Decreases	Decreases	No impact

- ◆ We will now consider two approaches to tuning the PID controller:
  1. Ziegler-Nichols method
  2. Trial-and-error manual tuning



# Ziegler-Nichols method of tuning PID controller

1. Set  $K_d$  and  $K_i$  to **zero**.
2. Adjust  $K_p$  from 0 until the system starts to **oscillate** at certain frequency.
3. **Measure** the value  $K_u = K_p$ , and the oscillation period as  $T_u$ .
4. **Set** the various **gain factors** according to the follow formula and table:

$$u(t) = K_p(e(t) + \frac{1}{T_i} \int_0^t e(\tau) d\tau + T_d \dot{e}(t))$$

<b>Control Type</b>	$K_p$	$T_i$	$T_d$
$P$	$0.5K_u$	-	-
$PI$	$0.45K_u$	$T_u/1.2$	-
$PD$	$0.8K_u$	-	$T_u/8$
<i>classic PID</i>	$0.6K_u$	$T_u/2$	$T_u/8$

# Manual method of tuning PID controller

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1. Set  $K_p$ ,  $K_d$  and  $K_i$  to zero.
  2. Start with a small  $K_p$ , double it each time until the system starts to oscillate.
  3. Half the value of  $K_p$ .
  4. Start with a small  $K_d$ , double it each time until the system starts to oscillate.
  5. Half the value of  $K_d$ .
  6. Start with a small  $K_i$ , double it each time until the system starts to oscillate.
  7. Half the value of  $K_i$ .
- ◆ If you are only using PD or PI controller, skip the irrelevant steps.
  - ◆ Fine tune the various gain until you get the response you want.

# Segway balancing with PID controller

- ◆ For the team project, we need to balance the Segway using feedback control.
- ◆ Instead of using motor speed as the control variable, we should use the **pitch angle** as the **control variable**.
- ◆ Available for us to use is the **pitch angle** (after passing through a **complementary filter**)  $p$  and the **rate of change of pitch angle** (from gyroscope alone)  $\dot{p}$ .
- ◆ Since we have  $\dot{p}$  available, the best controller to use is a PD controller, where the control variable is the pitch angle  $p$ .
- ◆ For **normal balancing** action, the **set-point  $r(t)$  is 0**, i.e. upright position.
- ◆ The controller output value  $w(t)$  is derived with all P, I and D terms of the pitch angle:

$$w(t) = K_p e(t) + K_d \dot{e}(t) + K_i \int e(\tau) d\tau$$

- ◆  $w(t)$  is used to drive the motors forward or backward in order to keep pitch angle  $p = 0$ , by producing PWM values to drive the two motors.
- ◆ To **move** the Segway forward or backward, **change the set-point  $r(t)$**  to some other values (a small positive or negative angle).
- ◆ To **turn** right or left, you need also to **adjust the ratio** of PWM duty cycle between the two motors.